

TMC2360

Video Output Processor VGA to NTSC/PAL

Features

- · Single-package graphics to video conversion
- Multiple input formats
 - 640x480 50/60 Hz, 800x600 50 Hz
- Multiple output standards
 - NTSC, NTSC-EIA, PAL-B/G/H/I, PAL-M
- · Composite and S-video output formats
- · No external memory required
- Horizontal and vertical positioning inputs
- Configuration set by 11 switches
 - Microcontroller port optional
- Auto sync polarity detection
- · Internal color bars
- 3-channel 8-bit input digitizer
- 2x oversampling 9-bit D/A converters
- Accepts programming through H and V timing
 VESA DPMS, filter modes, video standard
- Single +5V power supply

Applications

- VGA to video converter modules
- Computer video outputs
- · Video games
- · TV VGA in

Description

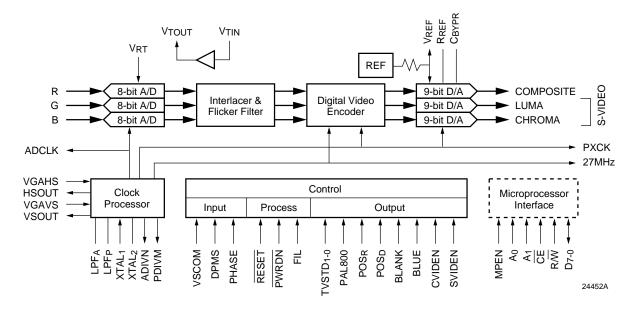
The TMC2360 converts RGB video and sync from a standard VGA source into NTSC or PAL video. Composite and S-Video outputs are compliant with SMPTE-170M and CCIR-656 specifications. A fully-integrated 3-line flicker filter provides three selectable operating modes.

Reference designs are available from the factory. System implementation requires an absolute minimum of external components. Critical timing is derived from a single 27 MHz crystal or an external clock reference.

All functions are directly controlled via package pins. Video and filtering modes may be programmed through either a microcontroller port or Vertical Sync Communications (VSCOM). VESA Display Power Management Signaling (DPMS) is supported.

Power is derived from a single +5V supply. Package is an 80-lead Metric Quad Flat Pack (MQFP), available in 2.0mm (KM) or 2.7mm (KL) package thickness.

Block Diagram



Functional Description

The TMC2360 is a VGA to Video converter capable of producing video signals conforming to NTSC and PAL standards using a single low-cost application circuit. Within the TMC2360 is all of the active circuitry required to generate a television signal with outstanding image quality in a standalone application.

Incoming VGA source signals must be 2X the frame and 2X the line rate of the outgoing TV standard within a $\pm 2\%$ tolerance. Supported VGA formats are 640x480 at 60 Hz for NTSC and PAL-M, and 640x480 and 800x600 at 50 Hz for PAL B/G/H/I.

The TMC2360 is ideal for portable converter applications, as well as integration into notebook and palmtop computers and video games.

Input Section

Analog VGA signals are digitized by three 8-bit A/D converters, operating at rates of up to 36 Ms/s. The signal range is 0 to 700 mV established by the reference voltage, VRT.

By connecting VTOUT to VRT, the A/D converter reference voltage may be supplied by an on-chip voltage follower with an input, VTIN, that may be varied from 0 to 2 volts to accommodate different input levels.

Clock Processor

Two phase-locked loops synthesize clocks from the VGA Horizontal Sync signal. One loop generates ADCLK, which is used internally as the A/D sample clock. A second PLL generates PXCK, which is used internally as the digital encoder clock.

Either internal or external phase-locked loops may be selected by programming pins A₁ and A₀. For internal loops, loop filters must be connected to LPFA and LPFP.

With external phase-locked loops, the internal divide by N and divide by M counters are still used. Only the phase detector, charge pump and VCO need be located in the external controller.

A stable timebase reference for subcarrier generation is derived from a 27 MHz crystal, or a TTL clock applied to pin XTAL₁.

To synchronize the video encoder, vertical timing is derived from VGAVS, the VGA vertical sync signal. VGAHS and VGAVS signals of either polarity are accepted.

VESA Display Power Management Signaling functions may be enabled with the DPMS pin. Using the DPMS protocol, operational commands may be communicated to the TMC2360 via VGAHS and VGAVS signals. DPMS STAND-BY or DPMS SUSPEND modes set the processor to sleep and blanks the screen. DPMS OFF sets the processor, A/Ds and D/As to sleep with a blanked screen.

Vertical Sync Communications may be enabled with the VSCOM pin to detect the number of VGAHS pulses during the VGAVS period. With VSCOM, the Flicker Filter mode and the Video Standard may be selected by commands communicated via the VGA sync signals.

Flicker Filter

Flicker may be traded-off against vertical resolution with a three-line adaptive flicker filter. A single toggle pin (FIL) selects either High Filter, Medium Filter, or No Filter modes. A fourth mode, Color Bars, is useful for video setup and as a reference point for filter selection.

Video Encoder

Unless VSCOM is enabled, TVSTD₁₋₀ pins select the TV standard to be either NTSC, PAL or PAL-M.

Relative to the bezel framed by horizontal and vertical sync, the image may be moved right/left, and down/up by pulsing the POS_R and POS_D pins. BLANK suppreses the image, setting the screen either black or blue according to the state of the BLUE pin.

NTSC (SMPTE 170M) and PAL (CCIR 624) video signals are produced by three 9-bit D/A converters that can drive the 37.5Ω load of a double-terminated 75Ω line. Digital 2X oversampling minimizes $\sin X/X$ distortion, facilitating use of low-cost output filters.

Composite and S-Video D/As are independently enabled via CVIDEN and SVIDEN pins to minimize power dissipation.

Control Processor

TMC2360 setup and control is derived from external switches and push buttons. Schmitt trigger inputs reject external noise. Unused controls may be preset by hardwiring inputs to ground or V_{CC}.

Encoder Output Current

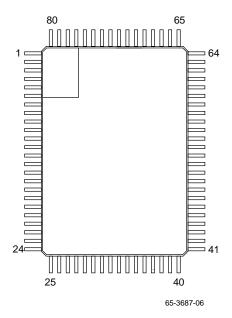
Output current is established by V_{REF} and an external resistor connected between R_{REF} and ground. An internal 1.235 volt reference is buffered from V_{REF} by a resistor, so V_{REF} may be overridden by an external voltage. Output current may be calibrated by resistor selection or setting a potentiometer attached to R_{REF} .

To minimize DAC noise, a bypass capacitor must be connected from CBYPR to an adjacent VDDA pin.

Microprocessor Port

Instead of utilizing control pin inputs, two operational modes, TV Standard and Flicker Filter, may be selected by writing to the VGA control register. Five registers may be read: address, VGA0, VGA1, Revision ID and Part ID.

Pin Assignments – 80-Lead MQFP Package (KL or KM)



Pin	Name	Pin	Name	Pin	Name	Pin	Name
1	AGND	21	D ₄	41	PAL800	61	DGND
2	COMPOSITE	22	D ₃	42	FIL	62	V_{DD}
3	R _{REF}	23	D ₂	43	DPMS	63	AGND
4	V _{REF}	24	D ₁	44	VSCOM	64	LPFA
5	DGND	25	D ₀	45	VGAVS	65	VDDPLLA
6	PHASE	26	VSOUT	46	VGAHS	66	AGND
7	V _{DD}	27	HSOUT	47	POSR	67	LPFp
8	DGND	28	DGND	48	POSD	68	VDDPLLP
9	V_{DD}	29	V_{DD}	49	AGND	69	XTAL ₂
10	DGND	30	MPEN	50	В	70	XTAL ₁
11	27MHZ	31	A ₁	51	V_{DDA}	71	RESET
12	PXCK	32	A ₀	52	V_{DDA}	72	PWRDN
13	ADCLK	33	R/W	53	G	73	SVIDEN
14	V _{DD}	34	CE	54	AGND	74	CVIDEN
15	DGND	35	V _{DD}	55	V _{RT}	75	V _{DDA}
16	ADIVN	36	DGND	56	V _{TOUT}	76	V_{DDA}
17	PDIVM	37	BLUE	57	V_{TIN}	77	CBYPR
18	D ₇	38	BLANK	58	AGND	78	CHROMA
19	D ₆	39	TVSTD ₁	59	R	79	AGND
20	D ₅	40	TVSTD ₀	60	V _{DDA}	80	LUMA

Pin Descriptions

		Type/	
Pin Name	Pin Number	Value	Pin Function Description
Clocks			
ADCLK	13	TTL	A/D Converter Clock Output. Generated by an internal phase-locked loop slaved to VGAHS.
ADIVN	16	TTL	Internal ADCLK divided by N Output. Enabled by A ₀ and A ₁ pins according to Table 1 or by VGA Control Register 1.
LPFA	64	_	A/D PLL Loop Filter Connection. An external RC network is connected here.
PXCK	12	TTL	Encoder Clock Output. Generated by an internal phase-locked loop slaved to VGAHS.
PDIVM	17	TTL	Internal PXCK divided by M Output. Enabled by A ₀ and A ₁ pins according to Table 1 or by VGA Control Register 1.
LPFp	67	_	Encoder PLL Loop Filter Connection. An external RC network is connected here.
XTAL ₁₋₂	70, 69	_	Subcarrier Reference Crystal/Clock. Connection terminals for an external 27 MHz crystal. Alternatively, the XTAL1 pin may be used as an input from an external oscillator or clock. Subcarrier frequency accuracy is based on this clock.
27MHZ	11	TTL	Subcarrier Reference Clock Output. Buffered TTL output from 27MHz crystal oscillator/reference clock.
VGAHS	46	CMOS _S	VGA Horizontal Sync. Incoming VGA sync may be of either polarity (active LOW or active HIGH). VGAHS frequency must be within 2% of the nominal specified value. The VGAHS pin has a light pull-up and a Schmitt trigger.

Pin Descriptions (continued)

Pin Name	Pin Number	Type/ Value	Pin Function Description	
VGAVS	45	CMOSs	VGA Vertical Sync. Incoming VGA sync may be of either polarity (active LOW or active HIGH). The VGAVS pin has a light pull-up and a Schmitt trigger.	
HSOUT	27	TTL	Buffered Horizontal Sync Output. Follows VGAHS.	
VSOUT	26	TTL	Buffered Vertical Sync Output. Follows VGAVS.	
Global Conti	rols			
TVSTD ₁₋₀	39, 40	CMOSP	Video Output Standard Select. Preprogrammed into the TMC236 are timing, subcarrier frequency and phase parameters corresponding to worldwide NTSC and PAL standards. TVSTD ₁₋₀ select one of four sets of parameters to set up the encoder. Frame rate of the graphics source must be twice the frame rate of the selected video standard.	
PAL800	41	CMOSP	Resolution Select for PAL. Sets number of samples per VGA line.	
DPMS	43	CMOSP	Display Power Management Signaling Enable. When HIGH, the operational state of the TMC2360 is controlled by the pulse activity on VGAHS and VGAVS. When LOW, the state of the TMC2360 is controlled only by input pins.	
FIL	42	CMOSs	Flicker Filter Mode Select. The adaptive flicker reduction filter may be configured for HIGH filtering, MEDIUM filtering, or NO filtering by pulsing this input HIGH. FIL defaults to HIGH-filter mode upon powerup. If VSCOM is HIGH, the filter mode will be selected by the pulsewidth of VGAVS, and FIL will be ignored. The FIL input is a Schmitt trigger.	
VSCOM	44	CMOSP	Vertical Sync Communications Enable. When HIGH, vertical sync pulse width (VGAVS) will control the filter mode, and FIL will be ignored.	
PWRDN	72	CMOSP		
RESET	71	CMOSP	Reset. Initializes internal registers.	
PHASE	6	CMOSP	Sampling Phase Control. Shifts the A/D sampling phase by 180°. Set Phase = LOW.	
Encoder Cor	ntrols			
CVIDEN	74	CMOSP	Composite Video D/A Power Enable. When HIGH, the COMPOSITE D/A converter is enabled. When LOW it is disabled to save power.	
SVIDEN	73	CMOSP	S-Video D/A Power Enable. When HIGH, the CHROMA and LUMA D/A converters are enabled. When LOW, they are disabled to save power.	
BLANK	38	CMOSP	Blank Screen Generator. When HIGH, the color selected by BLUE is displayed on the screen. When LOW, incoming video from the internal FIFO is encoded.	
BLUE	37	CMOSP	Blank Screen Color Selector. When HIGH, the screen will be blanked to blue when BLANK is HIGH. When LOW, the screen will be blanked to black when BLANK is HIGH.	

Pin Descriptions (continued)

Pin Name	Pin Number	Type/ Value	Pin Function Description
POSR, D	47, 48	CMOSs	TV Image Position Controls. Position controls shift the VGA image horizontally or vertically, revealing portions that are found near the edges or in the overscan areas. Default power-up position is the midpoint of the adjustment range. POSD,R inputs are Schmitt triggers.
A/D Converte	r Interface		
R, G, B	59, 53, 50	700 mV	Analog RGB Inputs. Analog red, blue and green inputs to the A/D converters from incoming VGA signals. Nominal voltage range is 0.0 to +700 mV.
VTIN	57	750 mV	A/D Converter Reference Buffered Input. Buffer is a voltage follower that may be connected to VRT.
Vтоит	56	750 mV	A/D Converter Reference Buffered Output . May be connected to V _{RT} to supply current to A/D converter reference resistors. In power down mode, V _{TOUT} drops to zero.
VRT	55	750 mV	A/D Converter Reference Input, Unbuffered. Supplies current to A/D converter reference resistors. May be driven from V _{TOUT} . Voltage range is 0.5 to 2.0 volts.
Video Outputs	S		
COMPOSITE	2	1 V p-p	NTSC/PAL Video Output, Composite Video. NTSC/PAL baseband composite output can drive 1 Volt p-p video into a 37.5 Ω load. Contains sync, subcarrier and active video information to drive monitors, projectors, VCRs, and other video devices.
LUMA	80	1 V p-p	Luminance-only Video. This analog monochrome video output can drive 1 Volt p-p video into a 37.5 Ω load. Contains all sync and active video information necessary to drive black-and-white video devices.
CHROMA	78	1 V p-p	Chrominance-only Video. This analog output can drive a 37.5 Ω load. CHROMA signal, when combined with LUMA comprises an S-Video signal suitable for driving monitors, projectors, VCRs, and other S-Video devices.
Voltage Refer	ence		
VREF	4	+1.23 V	Voltage Reference Input/Output. Output of an internal 1.23 Volt band-gap voltage reference. If unconnected, except for a 0.1F capacitor to ground for noise decoupling, the internal reference will be used for the three D/A converters. An externally generated voltage reference of +1.2 Volts applied to the VREF pin will override the internal voltage reference and become the new reference for the D/A converters.
Свург	77	0.1 μF	Reference Bypass Capacitor. An external 0.1F capacitor should be connected between CBYPR and VDDA to reduce noise on the internal reference circuitry.
RREF	3	392Ω	Current-setting Resistor. A 392Ω resistor connected between the RREF terminal and ground establishes the reference current for the three internal D/A converters. Resistor value determines the full-scale output current (and therefore the peak video level) of the D/A converters.

Pin Descriptions (continued)

Pin Name	Pin Number	Type/ Value	Pin Function Description
Microproces	sor Interface		
MPEN	30	TTL	Microprocessor Port Enable. With MPEN = LOW, the port is disabled and control is via individual pins. With MPEN = HIGH, selected functions may be accessed through the microprocessor port.
A ₀	32	TTL	External PXCK Phase-locked Loop Select/Address Bit. Dual function pin. If MPEN = HIGH, A ₀ is the microprocessor port address bit A ₀ . If MPEN = LOW, A ₀ configures the PXCK PLL inputs and outputs.
A1	31	TTL	External ADCLK Phase-locked Loop Select. If MPEN = LOW, A1 configures the ADCLK PLL inputs and outputs.
R/W	33	TTL	Read/Write. Read/Write selects the direction of the 8-bit data bus.
CE	34	TTL	Chip Enable. When $\overline{CE} = H$, D_{7-0} are high impedance. When $\overline{CE} = L$, R/\overline{W} sets D_{7-0} to either the read or write state.
D7-0	18, 19, 20, 21, 22, 23, 24, 25	TTL	Data bits. Data bus is high impedance unless $\overline{CE} = L$ and $R/\overline{W} = H$.
Power and C	Fround		
V _{DD}	7, 9, 14, 29, 35, 62,	+5.0 V	Digital Power Supply. Supplies +5V power to internal digital circuits.
VDDA	51, 52, 60, 75, 76	+5.0 V	Analog Power Supply. Supples +5V power to internal analog circuits. VDD and VDDA must originate from the same source.
VDDPLLA	65	+5.0 V	A/D Phase Locked Loop +5V Power. VDDPLLA and VDD must originate from the same source.
VDDPLLP	68	+5.0 V	Encoder Phase Locked Loop +5V Power. VDDPLLP and VDD must originate from the same source.
DGND	5, 8, 10, 15, 28, 36, 61	0.0 V	Digital Ground. Ground point for internal digital circuits.
AGND	1, 49, 54, 58, 63, 66, 79	0.0 V	Analog Ground. Ground point for internal analog circuits. DGND and AGND should be connected to the same ground plane.

Notes:

CMOSP = CMOS with light pull-up CMOSS = CMOS with Schmitt Trigger

6

Clocks

There are three internal clocks, ADCLK, PXCK, and 27MHz. ADCLK is the clock for the A/D converters with sampling on the edge selected by PHASE. PXCK is the encoder clock for sequencing data to the D/A converters at a 2X rate. 27MHz is the reference clock from which the chroma subcarrier data is synthesized.

ADCLK and PXCK clocks may be derived from internal or external phase-locked loops. Only an external controller containing the phase detector, charge pump, and VCO is needed for each loop. An internal divider programmed with the correct counts is included within the TMC2360.

Control of the internal/external PLL modes is either via the A_0 and A_1 pins (see Table 1) or through the VGA1 Control register, which can be programmed via the microprocessor port.

With control via A₀ and A₁ pins, the outputs PDIVM and ADIVN are high impedance unless either PLL is programmed to be external. A₀ and A₁ also control the direction of the ADCLK and PXCK clock pins. Each clock (ADCLK and PXCK) is an output in the internal mode and an output in the external mode.

Table 1. Internal/External Phase-locked Loop Selection (MPEN = L)

A1	A ₀	PLLA	PLLP	ADIVN	PDIVM	ADCLK	PXCK
0	0	Internal	Internal	High-Z	High-Z	Output	Output
0	1	Internal	External	1/N	1/M	Output	Input
1	0	External	Internal	1/N	1/M	Input	Output
1	1	External	External	1/N	1/M	Input	Input

A/D Clock (ADCLK)

ADCLK is the buffered analog-to-digital converter clock output which is derived from incoming VGA horizontal sync (VGAHS) by a phase-lock loop (PLL). Either an internal PLL or an external PLL controller may be selected by programming pin A_1 .

A/D Clock frequency is set by the PLL divide-by-N counter where N is the number of A/D samples between the horizontal sync pulses in each VGA line.

Pre-programmed values selected by the TVSTD₁₋₀ and PAL800 control inputs set N and the clock frequency as shown in Table 2.

Table 2. VGA A/D Clock

Television Standard	TVSTD ₁₋₀	PAL800	ADCLK Freq. (MHz)	N
NTSC	0x	0	25.175	800
PAL640	10	0	25.250	808
PAL800	10	1	36.000	1152
PALM	11	0	25.175	800

For an internal loop, the recommended loop filter to be connected to the LPF_A pin is shown in Figure 1.

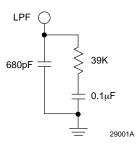


Figure 1. PLL Low Pass Filter

Pixel Clock (PXCK)

PXCK is the buffered video encoder clock output which is derived from the incoming VGA HSYNC signal by a second phase-lock loop. Either an internal PLL or an external PLL controller may be selected by programming pin A₀.

With the internal loop selected, Figure 1 shows the recommended loop filter, which should be connected to LPFP.

Clock frequency is set by the PLL divide-by-M counter. M is the number of encoder samples between horizontal sync pulses. Because of 2X oversampling, the pixel rate is twice the TV square pixel rate. Table 3 shows the video output clock rates and M values.

Table 3	NTSC	and PAL	Pival	Clacks
Table 5.	14 1 20	and FAL	CIXEL	CHURA

Television Standard	Line Rate (kHz)	Pixel Rate (MHz)	М
NTSC	15.734	24.545	780
PAL	15.625	29.500	944
PALM	15.750	24.570	780

Reference Clock

Accuracy of the PAL/NTSC subcarrier depends on the 27 MHz reference signal applied to the XTAL1. This signal may be derived from a clock connected directly to the XTAL1 pin or a crystal and capacitors connected across XTAL1-2 as shown in Figure 2.

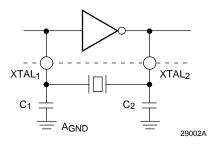


Figure 2. Crystal Oscillator Circuit

Capacitors C₁ and C₂ must be adjusted to trim the frequency within 20 ppm. C₁ and C₂ have the same value, with series capacitance equal to the load recommended for the crystal.

Typical crystal parameters are 50 ppm accuracy with a 20 pF load and ±80 ppm pullability.

Digitizing

A/D Reference

An on-chip voltage follower is included to supply current to the ADC reference VRT from VTOUT. A stable voltage source greater than the input peak amplitude should be connected to VTIN.

Input Signal Conditioning

ADC performance can be optimized by driving the RGB video inputs with either a 75 ohm or a low impedance source.

Input Format Selection

One of four input VGA formats can be accepted by setting the TVSTD₁₋₀ and PAL800 inputs as shown in Table 4. Each VGA input option corresponds to a VGA active video area, frame rate and line rate with a corresponding TV output format.

Table 4. VGA Input Formats

TVSTD ₁₋₀	PAL800	H x V Input Pixels	Frame Rate (Hz)	Line Rate (kHz)
0x	0	640 x 480	59.94	31.469
10	0	640 x 480	50	31.250
10	1	800 x 600	50	31.250
11	0	640 x 480	60	31.469

Processing

Flicker Filter

Annoying artifacts can be eliminated by selecting one of three filter modes, which trade-off vertical resolution against flicker. Without the filter, one contrasting VGA line may be encoded into one field of the TV video, which will flicker at 30 Hz with NTSC and 25 Hz with PAL.

As shown in Table 5, if VSCOM = LOW, pulsing the FIL pin indexes the TMC2360 through a loop of three filter modes and a color bar pattern.

Table 5. FIL Filter Mode Select Sequence

FIL	Filter Mode
↓ •	HIGH (default)
↓	MEDIUM
↓	No filter
	Color bars

Video Encoder

D/A Reference

Peak D/A converter current for the video outputs is set by a resistor connected to R_{REF} . For 1 volt video, with a 37.5 ohm load, the correct value of R_{REF} is 392 ohm. To trim the video output level, R_{REF} can be replaced with a potentiometer. (See the Applications Circuit, Figure 15).

Television Standard Selection

NTSC and PAL standards are preprogrammed into the TMC2360 to preset horizontal and vertical timing, subcarrier frequency, and chrominance phase. Frame rate of the VGA source must match the field rate of the selected video standard.

Depending upon the status of the Vertical Sync Communications (VSCOM) pin, the video output format may be selected by either the TVSTD1-0 pins or by VSCOM codes.

Table 6 shows how the $TVSTD_{1-0}$ pins select the TV output format with VSCOM = LOW.

Table 6. TVSTD Control When Under Manual Control (VSCOM = 0)

TVSTD ₁₋₀	Television Standard	Video Field Rate
01	NTSC	59.94 Hz
00	NTSC-EIA	59.94 Hz
10	PAL/B, G, I	50 Hz
11	PAL/M	60 Hz

If VCSOM = HIGH, Vertical Sync Communications are enabled. VSCOM is described under the Vertical Sync Command section below.

Image Positioning

POS_D and POS_R position controls change encoder timing relative to incoming PC video, shifting the viewed image either horizontally or vertically to reveal portions of the image located near the edges or in the overscan areas. At power-up, the default position is the midpoint of the adjustment range.

Each POSD HIGH pulse moves the TV window down eight lines. At the lowest position (-64 lines) the direction reverses and the pulses move the image up in 8-line increments to the highest position (+64 lines). At this point the direction again reverses and the next sixteen pulses move the image down to the lowest position.

Each POSR HIGH pulse moves the TV window eight pixels to the right. At the maximum right position (+64 pixels) the direction reverses and the next sixteen pulses move the window left to the maximum left position (-64 pixels). Direction again reverses and the next sixteen pulse move the image right.

Without a RESET, POSD,R controls will loop, causing the window to move down/up, right/left.

Blank and Blue

TV video can set to active (converted VGA source), blank or blue by the BLANK and BLUE inputs.

Table 7. Video Output for BLANK and BLUE inputs

BLANK	BLUE	Video
L	X	Video
Н	L	Black
Н	Н	Blue

Power Management

D/A Power Control

Table 8 shows how the Composite and S-video outputs are enabled by the CVIDEN and SVIDEN controls. If

DPMS = LOW, the display will be blue if either or both VGAHS or VGAVS are inactive.

Table 8. Video Output Control

CVIDEN	SVIDEN	VGAHS & VGAVS	Compos- ite Video	S-Video
Н	Х	YES	Active	X
Х	Н	YES	Χ	Active
L	L	Х	Blank	Blank
Н	Н	NO	Blue	Blue

Powerdown Mode

PWRDN eliminates current drain by the D/A converter, VTOUT voltage follower and clock outputs. With PWRDN = HIGH, all outputs are enabled. If PWRDN = LOW, all outputs are disabled including the ADCLK, PXCK, VTOUT and the D/A converters.

Software Control

Display Power Management Signaling (DPMS)

Display Power Management is compliant with VESA DPMS Proposal 1.0. With DPMS = HIGH, the operational state of the TMC2360 is controlled by the pulse activity on VGAHS and VGAVS.

Table 9 shows how the TMC2360 responds. "No Pulses" on VSYNC or HSYNC is declared on the second missing VSYNC or HSYNC pulse. Following an OFF state, detection of VGAHS and/or VGAVS restores either the Suspend, Standby, or On state.

Regardless of the state of DPMS, absence of VGAHS and/or VGAVS pulses will cause the TMC2360 processor to sleep. However, with DPMS = HIGH, in the Off state, the A/Ds, D/As, and clocks are also set to sleep.

Vertical Sync Communications (VSCOM)

VSCOM is a unique feature incorporated into the TMC2360. With VSCOM = HIGH, the TMC2360 interprets commands that are encoded within the VGA vertical sync period. TV Standard and Flicker Filter may be selected by commands from the PC that supplies the VGA signal.

During the vertical sync period, the number of horizontal sync pulses is counted. Table 10 shows how the selected Television Standard changes with the code set into the TVSTD₁₋₀ inputs and the number of horizontal syncs per vertical sync interval.

Table 11 shows how the Filter Mode is selected with VSCOM. Notice that if one standard, such as PAL800 is selected, then counts of 10, 11 and 12 enable selection of three filter modes.

TMC2360	PRODUCT SPECIFICATION
---------	-----------------------

State	VGAHS	VGAVS	TMC2360 State
On	Pulses	Pulses	On, video active
Stand-by	No Pulses	Pulses	Stand-by, screen blanked (color set by BLUE)
Suspend	Pulses	No Pulses	Stand-by, screen blanked (color set by BLUE)
Off	No Pulses	No Pulses	Off, screen black (equivalent to PWRDN, except A/D Clock and 27MHZ oscillator are active.)

Table 10. TVSTD Control When Under Software Control (VSCOM = 1)

TVSTD ₁₋₀	Hsyncs per Vsync Interval	Television Standard	VGA Frame Rate
XX	10,11,12	PAL 800 (B, G, I)	50 Hz
X0	6,7,9	PAL 640 (B, G, I)	50 Hz
01	6,7,9	PAL-M	60 Hz
11	6,7,9	(reserved)	
0X	1-5,8,14,15	NTSC-EIAJ	60 Hz
10	1-5,8,14,15	NTSC	60 Hz
11	1-5,8,14,15	(reserved)	
XX	0,13	Blank (with prior standard)	50/60 Hz

Table 11. VSCOM Filter Command Interpretation (VSCOM=1)

HSYNCs per VSYNC Interval	Filter Mode
5, 9, 12	No Filter
4, 7, 11	2-line filter
1, 2, 3, 6, 8, 10, 14, 15	3-line filter
0, 13	Blank (with prior mode)

Microprocessor Interface

Table 12 shows the five registers that may be read after the Address Register has been loaded. Only the two VGA registers and the Address Register may be written. For microprocessor control, 40h must be written into VGA register 1. To disable control through VGA Register 0, 00h must be written into register 1. Bit assignments of VGA register 0 are listed in Table 13. Bit assignments of VGA register 1 are listed in Table 14.

For communications through the microprocessor port, set MPEN = HIGH. Read and write timing is shown in Figure 3 and Figure 4. To access a VGA register, write the address of the selected register into the lower four bits of the eight bit Address Register with $A_0 = LOW$. Then with $A_0 = HIGH$, read or write to the selected register.

Table 12. Microprocessor Addressable Registers

A1	Α0	Address Register Value	Register
0	0	x	Address Register
0	1	x0	VGA Register 0
0	1	x1	VGA Register 1
0	1	x6	Revision ID (xxh)
0	1	x7	Part ID (0Fh)

Table 13. VGA Control Register 0 Bit Map

Bit #	Name	Function		
7	_	0 (reserved)		
6-5	FIL	0 0 No Filter		
		0 1 2-line Filter		
		1 x 3-line Filter		
4-2	TVSTD	000 NTSC-EIA		
		001 NTSC		
		010 PALM		
		0 1 1 PAL640		
		1xx PAL800		
1-0	VIDEO	0 0 Black		
		01 Ramp		
		1 0 Color Bars		
		1 1 Normal		

Table 14. VGA Control Register 1 Bit Map

Bit #	Name	Function
7	_	0 (reserved)
6	MODE	0 Normal
		Bypass FIL and TVSTD Pins
5-3	_	0 (reserved)
2	DIVMN	0 (reserved)
		1 Enable ADIVN and PVIDM
1	EXTPLL	0 Disable external PXCK PLL
		1 Enable external PXCK PLL
0	EXTPLLA	0 Disable external ACLK PLL
		1 Enable external ACLK PLL

Video Formats

Incoming VGA Formats

Table 15 and Table 16 show expected VGA Video input formats.

Outgoing TV Formats

Table 16 shows the four different TV formats that may be selected as outputs by the TVSTD₁₋₀ inputs.

Table 17 and Table 18 show the Horizontal and Vertical Timing for the NTSC and PAL formats.

Table 15. VGA Horizontal Timing Parameters

Television Standard	TVSTD1-0	PAL800	Line Rate (kHz)	Front Porch (pixels)	Horiz Sync (pixels)	Back Porch (pixels)	Active Video (pixels)
NTSC(-EIA)	0x	0	31.469	18	96	46	640
PAL/B, G, I	10	0	31.250	18	96	54	640
PAL/B, G, I	10	1	31.250	98	96	158	800
PAL/M	11	0	31.500	18	96	46	640

Table 16. VGA Vertical Timing Parameters

Television Standard	TVSTD1-0	PAL800	Frame Rate (Hz)	Line Rate (kHz)	Full Frame (lines)	Front Porch (lines)	Vert. Sync (lines)	Vsync + Back Porch (lines)	Active Video (lines)
NTSC(-EIA)	0x	X	59.94	31.469	525	13	2-16	32	480
PAL/B, G, I	10	0	50	31.250	625	61	2-16	84	480
PAL/B, G, I	10	1	50	31.250	625	1	2-16	24	600
PAL/M	11	х	60	31.469	525	13	2-16	32	480

Table 17. NTSC and PAL Horizontal Timing

				•						
Television Standard	Field Rate (Hz)	Lines per frame	Line Rate (kHz)	2x pix Rate (MHz)	fSC Freq. (MHz)	Front Porch pixels	Horiz Sync pixels	Back Porch pixels	Active Video pixels	Line H pixels
NTSC	59.94	525	15.734	24.545	3.579	18	58	58	646	780
PAL	50.00	625	15.625	29.500	4.433	18	70	82	774	944
PALM	60	525	15.734	25.570	3.576	18	58	58	646	780

Table 18. NTSC and PAL Vertical Timing

TV Std	Field Rate (Hz)	Lines per frame	Line Rate (kHz)	Front Porch (lines)	Vertical Sync (lines)	Back Porch (lines)	Active Video (lines)
NTSC	59.94	525	15.734	3-3.5	3	14-14.5	242.5
PAL	50.00	625	15.625	2.5	2.5	21	286.5
PALM	60.00	525	15.750	3	3	14	242.5

Timing Diagrams

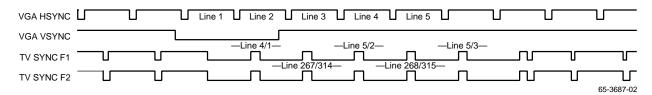


Figure 3. Nominal VGA-NTSC/PAL Vertical Sync Timing

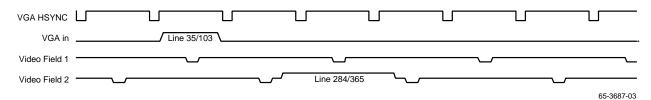


Figure 4. Nominal VGA640-NTSC/PAL, No Filter Timing

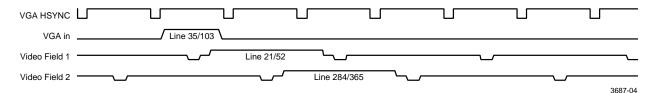


Figure 5. Nominal VGA-NTSC/PAL, Medium Filter Timing

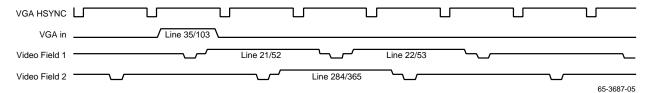


Figure 6. Nominal VGA-NTSC/PAL, High Filter Timing

Timing Diagrams (continued)

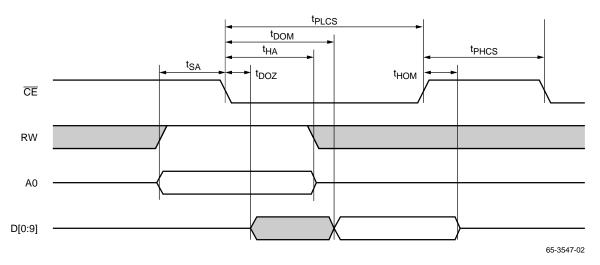


Figure 7. Microprocessor Read Timing

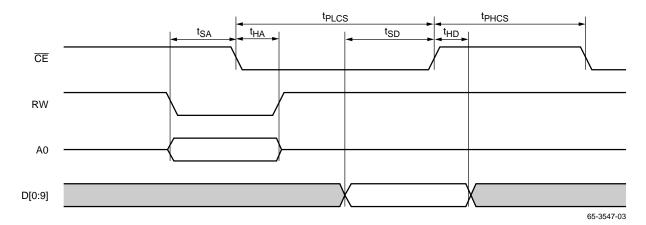


Figure 8. Microprocessor Write Timing

Equivalent Circuits

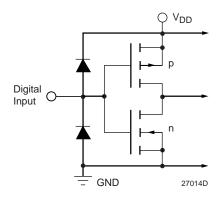


Figure 9. Equivalent Digital Input Circuit

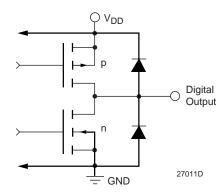


Figure 10. Equivalent Digital Output Circuit

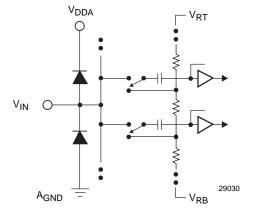


Figure 11. Equivalent A/D Input Circuit

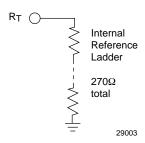


Figure 12. Equivalent A/D Reference Input Circuit

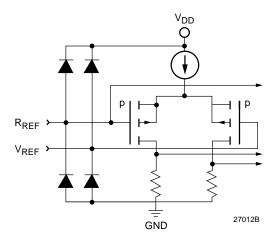


Figure 13. Equivalent D/A Reference Input Circuit

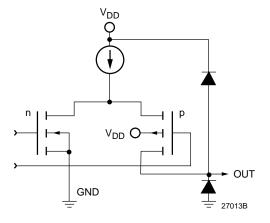


Figure 14. Equivalent D/A Output Circuit

Absolute Maximum Ratings (beyond which the device may be damaged)¹

Parameter	Min.	Тур.	Max.	Unit
Power Supply Voltages				
VDDA (Measured to AGND)	-0.5		7.0	V
V _{DD} (Measured to DGND)	-0.5		7.0	V
VDDA (Measured to VDD)	-0.5		0.5	V
AGND (Measured to DGND)	-0.5		0.5	V
Digital Inputs				
Applied Voltage (Measured to DGND) ²	-0.5		V _{DD} + 0.5	V
Forced current ^{3, 4}	-10.0		10.0	mA
Analog Inputs		•		
Applied Voltage (Measured to AGND) ²	-0.5		VDDA + 0.5	V
Forced current ^{3, 4}	-10.0		10.0	mA
Digital Outputs	•	!		
Applied Voltage (Measured to DGND) ²	-0.5		VDD + 0.5	V
Forced current ^{3, 4}	-6.0		6.0	mA
Short circuit duration (single output in HIGH state to ground)			1	second
Temperature				
Operating, Ambient	-20		110	°C
Junction			150	°C
Lead Soldering (10 seconds)			300	°C
Vapor Phase Soldering (1 minute)			220	°C
Storage	-65		150	°C
Electrostatic Discharge ⁵			±150	V

Notes:

- 1. Functional operation under any of these conditions is NOT implied. Performance and reliability are guaranteed only if Operating Conditions are not exceeded.
- 2. Applied voltage must be current limited to specified range.
- 3. Forcing voltage must be limited to specified range.
- 4. Current is specified as conventional current flowing into the device.
- 5. EIAJ test method.

Operating Conditions

Parameter		Min.	Nom.	Max.	Units
VDD	Digital Power Supply Voltage	4.75	5.0	5.25	V
VDDA	Analog Power Supply Voltage	4.75	5.0	5.25	V
AGND	Analog Ground (Measured to D _{GND})	-0.1	0	0.1	V
VRT	Reference Voltage, Top	0.5	0.75	2.0	V
V _{IN}	Analog Input Range	0		V _{RT}	V
VREF	External Reference Voltage		1.235		V
IREF	D/A Converter Reference Current (IREF = VREF/RREF, flowing out of the RREF pin)		3.15		mA
RREF	Reference Resistor, VREF = Nom		392		Ω
Rout	DAC Total Output Load Resistance		37.5		Ω
TA	Ambient Temperature, Still Air	0		70	°C

Electrical Characteristics

Parameter		Conditions	Min.	Тур.	Max.	Unit
Power Sup	oply Currents		•			
IDD	Operating	CVIDEN = H, SVIDEN = H		300	400	mA
IDDS	Standby	CVIDEN = L, SVIDEN = L		100		mA
IDDQ	Power-Down	PWRDN = L			5	mA
IDDSV	S-Video Active	CVIDEN = L, SVIDEN = H		250	350	mA
IDDCV	Composite Video Active	CVIDEN = H, SVIDEN = L		200	270	mA
Digital Inp	uts and Outputs					
Cı	Input Capacitance			5	10	pF
Co	Output Capacitance			10		pF
lін	Input Current, HIGH	VDD = Max, VIN = VDD			±10	μΑ
IIL	Input Current, LOW	V _{DD} = Max, V _{IN} = 0 V			±10	μΑ
VIHTTL	Input Voltage, Logic HIGH (TTL)		2.0			V
VILTTL	Input Voltage, Logic LOW (TTL)				0.8	V
VIHCMOS	Input Voltage, Logic HIGH (CMOS)		3.5			V
VILCMOS	Input Voltage, Logic LOW (CMOS)				1.5	V
VT+	Schmitt Trigger Positive Threshold			3.0		V
VT-	Schmitt Trigger Negative Threshold			0.8		V
Іон	Output Current, Logic HIGH				-2.0	mA
loL	Output Current, Logic LOW				2.0	mA
Voн	Output Voltage, HIGH	IOH = -2mA	2.4			V
VoL	Output Voltage, LOW	I _{OL} = 2mA			0.4	V
Analog In	puts			!		
CAI	A/D Input Capacitance	ADCLK = LOW ADCLK = HIGH		4 12		pF pF
RIN	A/D Input Resistance		500	1000		ΚΩ
ICB	A/D Input Current				±15	μА
VRO	Voltage Reference Output	Internal Reference	0.988	1.235	1.482	V
IRO	VREF Output Current	External VREF	-150		+150	μА
Analog O	utputs		•			
Voc	Video Output Compliance		-0.4		2	V
Rout	Video Output Resistance			15		ΚΩ
Соит	Video Output Capacitance	C _{OUT} = 0 mA, Frequency = 1 MHz		15		pF
los	Short-Circuit Current		-20		-80	mA
IILP	Input Current, LOW with pull-up		-100			μΑ

Switching Characteristics

Parameter		Conditions	Min.	Typ. ¹	Max.	Unit
Clocks		'		!	!	ļ.
fXTAL	Crystal/Reference Clock Frequency			27		MHz
fxTOL	Crystal/Reference Clock Frequency Tolerance			50		ppm
tPWH	Reference Clock Pulse Width, HIGH			18.5		ns
tpwL	Reference Clock Pulse Width, LOW			18.5		ns
Syncs						
fH	VGAHS Frequency	60 Hz Modes	30.840	31.469	32.100	KHz
		50 Hz Modes	30.630	31.250	31.880	KHz
NH	Lines per VGA frame	60 Hz Modes		525		
		50 Hz Modes		625		
		Tolerance			±0	
tpwhs	VGAHS Pulsewidth		2			μS
tvs-Hs	VGAVS to VGAHS Delay		0			ns
tDS	Sync Delay (VGA Sync to Sync Out)			100		ns
Video C	Output		•		•	•
tDOV	Analog Output Delay (PXCK Out to Video Out)				15	ns
tR	D/A Output Current Risetime (10% to 90%)			2.2		ns
tF	D/A Output Current Falltime (90% to 10%)			2		ns
SKEW	D/A to D/A Skew		-3	0	3	ns
Control	s	•	•			
tPWH	Control Input Pulse Width, HIGH			50		ns
tpwL	Control Input Pulse Width, LOW			50		ns
Micropi	ocessor Interface					
tPLCS	CS Pulse Width, LOW		12			ns
tPHCS	CS Pulse Width, HIGH		8			ns
tsa	Address Setup Time		2			ns
tHA	Address Hold Time		4			ns
tsD	Data Setup Time		7			ns
tHD	Data Hold Time		2			ns
tDOZ	Output Delay, CS to low-Z		4			ns
tHOM	Output Hold Time, CS to high-Z		9			ns
tDOM	Output Delay, CS to Data Valid				40	ns

^{1.} Values shown in Typ column are typical for $V_{DD} = V_{DDA} = +5V$ and $T_A = 25^{\circ}C$.

System Performance Characteristics

Paramete	er	Conditions	Min	Typ ¹	Max	Unit
A/D Converter Input			!	!	!	!
ELI	A/D Integral Linearity Error, Independent	VRT = 0.7V		±1		LSB
ELD	A/D Differential Linearity Error	VRT = 0.7V		±1		LSB
EAP	Aperture Error			30		ps
Еот	Offset Voltage, Top	RT – VIN for most positive code transition	-20	45	80	mV
Еов	Offset Voltage, Bottom	VIN for most negative code transition	30	65	110	mV
D/A Conv	verter Output			•		•
RES	D/A Converter Resolution		9	9	9	Bits
dp	Differential Phase	PXCK = 27 MHz, 40 IRE Ramp		0.5		degree
dg	Differential Gain	PXCK = 27 MHz, 40 IRE Ramp		1.5		%
PSRR	Power Supply Rejection Ratio	C _{BYP} = 0.1 μF, f = 1 KHz		0.5		%/ %VDD

^{1.} Values shown in Typ column are typical for $V_{DD} = V_{DDA} = +5V$ and $T_A = 25^{\circ}C$.

Applications Discussion

Applications Circuit

Two applications circuits are shown. Figure 15 is the standard applications circuit. Figure 16 is a simplified circuit that exploits many of the internal features of the TMC2360.

The standard applications circuit is intended for third party add-on modules that interface a PC to a TV. Emitter followers buffer the VGA RGB signals so that VGA monitor disconnect does not impact the level of the TV image. Reference clock source is a crystal which requires tuning. A clock oscillator is recommended. VREF and VTIN are derived from an external voltage reference with an adjustment potentiometer.

Although intended for VGA controller applications, the simplified circuit embodies many techniques that can be exported to other applications. Bandwidth of the incoming RGB signals is limited to 12 MHz to equalize intensities of parallel vertical lines. Pull-up resistors compensate for the bottom offset voltages of the A/D converters. RREF is comprised of two series resistors which divide down the voltage for VT. Sinx/x filters are omitted since with 2x oversampling, losses are small (~ 1 dB). A low pass equiphase filter in the CVBS channel cuts HF transients for vectorscope measurements. PLL power is derived from two separately filtered sources.

Grounding

Analog and digital circuits are separated within the TMC2360. To keep digital system noise from the A/D and D/A converters, it is recommended that power supply voltages (V_{DD} and V_{DDA}) originate from the same low-noise source, and that ground connections (DGND and AGND) be made to the analog ground plane. Power supply connections should be individually decoupled at each pin. Digital circuitry deriving input from the TMC2360 should be referred to the system digital ground plane.

Printed Circuit Board Layout

Designing with high-performance mixed-signal circuits demands printed circuits with ground planes. Overall system performance is strongly influenced by the board layout. Capacitive coupling from digital to analog circuits may result in poor A/D conversion. Consider the following suggestions for layout:

- Keep the critical analog traces as short as possible and as far as possible from all digital signals. Locate the TMC2360 near the board edge, close to the analog input/output connectors.
- 2. The power plane for the TMC2360 should be separate from that which supplies the rest of the digital circuitry. A single power plane should be used for all of the VDD pins. If the power supply for the TMC2360 is the same as that of the system's digital circuitry, power to the TMC2360 should be decoupled with ferrite beads and 0.1µF capacitors to reduce noise.
- The ground plane should be solid, not cross-hatched.
 Connections to the ground plane should have very short leads.
- 4. Decoupling capacitors should be applied liberally to VDD pins. Remember that not all power supply pins are created equal. They supply different circuits on the integrated circuit, each of which generate varying amounts and types of noise. For best results, use 0.1μF ceramic capacitors. Lead lengths should be minimized. Ceramic chip capacitors are the best choice.
- 5. If the digital power supply has a dedicated power plane layer, it should not be placed under the TMC2360, the voltage reference, or the analog inputs. Capacitive coupling of digital power supply noise from this layer to the TMC2360 and its related analog circuitry can have an adverse effect on performance.

The 27 MHz clock reference or crystal should be handled carefully. Jitter and noise on this clock will degrade performance. With an external clock, the line should be terminated to eliminate overshoot and ringing.

Locate phase locked loop components close to the relevant TMC2360 pins. Isolate these components from noise.

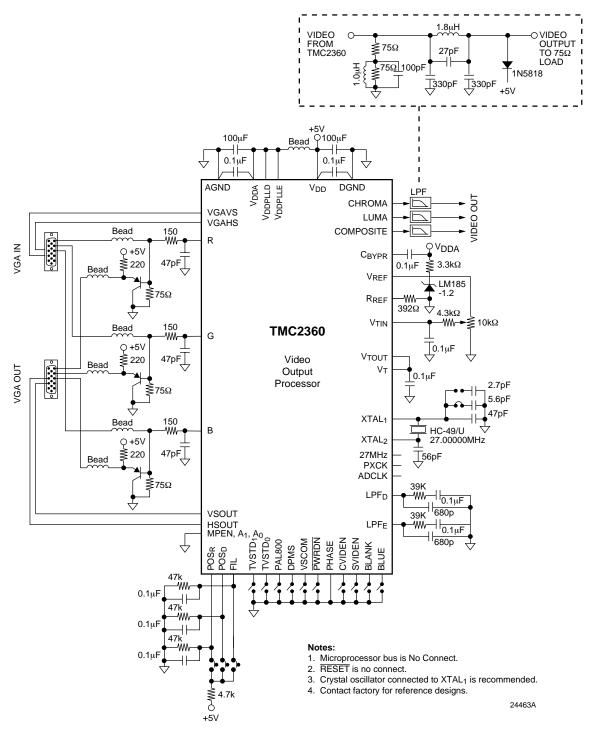


Figure 15. Typical application circuit with internal phase-locked loops

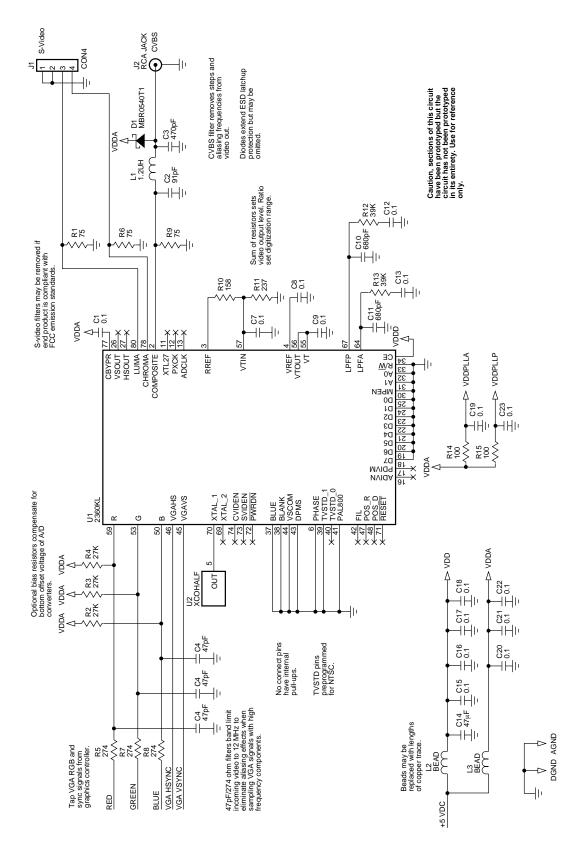


Figure 16. Simplified Applications Circuit

Suggested User Instructions

For a product incorporating the TMC2360, part of the documentation is expected to include operating instructions describing the functions of the TMC2360 user controls. A recommended text fragment follows.

Three external controls may be used to:

- position the image within the boundaries of the TV screen.
- 2) control vertical resolution.

Each control is a button that can be momentarily depressed to execute one cycle of the selected function.

Position Controls

Position controls shift the viewed image horizontally or vertically to reveal portions of the image that are located near the edges or in the overscan areas. At power-up, the default position is the midpoint of the adjustment range.

Each time the VERTICAL POSITION button is depressed, the TV window moves down by eight lines. At the lowest position (-64 lines) the direction reverses and depressing the button moves the image up in 8-line increments to the highest position (+64 lines). At this point the direction again reverses and the next sixteen pulses move the image down to the lowest position.

When the HORIZONTAL POSITION button is depressed, the TV window moves eight pixels to the right. At the maximum right position (+64 pixels) the direction reverses and the next sixteen pulses move the window left to the maximum left position (-64 pixels). Direction again reverses and the next sixteen pulses move the image right.

Flicker Filter

Annoying artifacts can be eliminated by selecting one of three filter modes which trade-off vertical resolution against flicker. For example, without the filter, a single VGA line is encoded into only one field of the TV display. If there is high contrast between this line and adjacent lines, it will flicker at 30 Hz in NTSC or 25 Hz in PAL.

Depressing the FILTER button indexes through the filter functions shown in the table below:

FIL	Filter Mode
↓ •	Soft vertical resolution without flicker
↓	Medium vertical resolution, some flicker
↓	Sharp vertical resolution with flicker
	Color bars

The filter should be selected for best appearance on the TV screen. Optimal selection will depend on the image being encoded.

Related Products

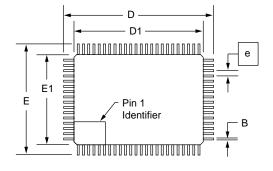
• TMC2302 Image Manipulation Sequencer

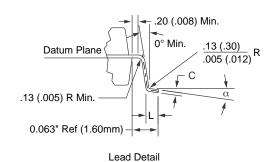
Mechanical Dimensions

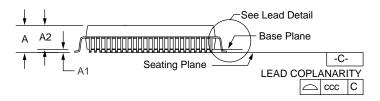
80-Lead MQFP Package (KL)

Symbol	Inches		Millin	Notes	
Symbol	Min.	Max.	Min.	Max.	Notes
Α	_	.134	_	3.40	
A1	.010	_	.25	_	
A2	.100	.120	2.55	3.05	
В	.012	.018	.30	.45	3, 5
С	.005	.009	.13	.23	5
D	.904	.923	22.95	23.45	
D1	.783	.791	19.90	20.10	
Е	.667	.687	16.95	17.45	
E1	.547	.555	13.90	14.10	
е	.0315	BSC	.80	BSC	
L	.025	.041	.65	1.03	4
N	8	0	8	0	
ND	2	:4	2	4	
NE	1	6	1	6	
α	0°	7°	0°	7°	
CCC	_	.004	_	0.10	

- 1. All dimensions and tolerances conform to ANSI Y14.5M-1982.
- 2. Controlling dimension is millimeters.
- Dimension "B" does not include dambar protrusion. Allowable dambar protrusion shall be .08mm (.003in.) maximum in excess of the "B" dimension. Dambar cannot be located on the lower radius or the foot.
- 4. "L" is the length of terminal for soldering to a substrate.
- 5. "B" & "C" includes lead finish thickness.





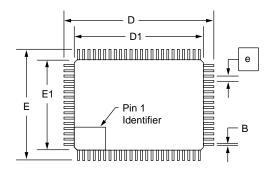


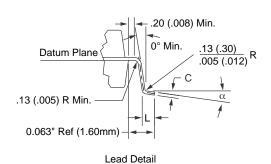
Mechanical Dimensions (continued)

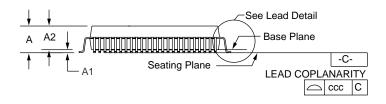
80-Lead MQFP Package (KM) - 3.2mm Footprint, 2.0mm Thickness

Cumbal	Inches		Millim	Notes	
Symbol	Min.	Max.	Min.	Max.	Notes
A	_	.096	_	2.45	
A1	.010	_	.25	_	
A2	.075	.083	1.90	2.10	
В	.012	.018	.30	.45	3, 5
С	.005	.009	.13	.23	5
D	.904	.923	22.95	23.45	
D1	.783	.791	19.90	20.10	
Е	.667	.687	16.95	17.45	
E1	.547	.555	13.90	14.10	
е	.0315	BSC	.80	BSC	
L	.029	.041	.73	1.03	4
N	8	0	8	0	
ND	2	4	24		
NE	1	6	16		
α	0°	7°	0°	7°	
CCC	_	.004	_	0.10	

- 1. All dimensions and tolerances conform to ANSI Y14.5M-1982.
- 2. Controlling dimension is millimeters.
- 3. Dimension "B" does not include dambar protrusion. Allowable dambar protrusion shall be .08mm (.003in.) maximum in excess of the "B" dimension. Dambar cannot be located on the lower radius or the foot.
- 4. "L" is the length of terminal for soldering to a substrate.
- 5. "B" & "C" includes lead finish thickness.







Ordering Information

Product Number	Temperature Range	Screening	Package	Package Marking
TMC2360KLC	0°C to 70°C	Commercial	80 Lead MQFP	2360KLC
TMC2360KMC	0°C to 70°C	Commercial	80 Lead MQFP	2360KMC

LIFE SUPPORT POLICY

FAIRCHILD'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS WRITTEN APPROVAL OF THE PRESIDENT OF FAIRCHILD SEMICONDUCTOR CORPORATION. As used herein:

- Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body, or (b) support or sustain life, and (c) whose failure to perform when properly used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury of the user.
- A critical component in any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

www.fairchildsemi.com